

PID stabilizáció

IX.28. p.
3.h

Ideális PID típusú stabilizátor:

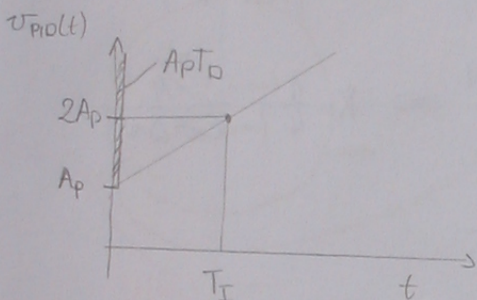
$$W_{PID}(s) = A_p \left(1 + \frac{1}{s T_I} + s T_D \right) = A_p + \frac{A_p}{T_I} \frac{1}{s} + A_p T_D s =$$

↑
integrátor diff.

$$W_{PID}(s) = k_p + k_I \frac{1}{s} + k_D \cdot s$$

$$V_{PID}(s) = W_{PID}(s) \cdot \frac{1}{s} = \frac{A_p}{s} + \frac{A_p}{T_I} \cdot \frac{1}{s^2} + A_p T_D$$

$$\mathcal{L}^{-1} V_{PID}(t) = A_p + \frac{A_p}{T_I} t + A_p T_D \delta(t) \quad ; \quad t \geq 0$$



Stabilizáció fajtái:

- P
- PI
- PD
- PID
- ~~-ID~~

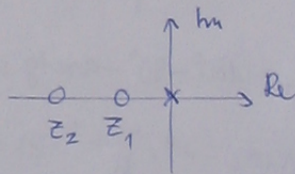
Agyórk alatt a gyökörrel van való's van!

Atrifeli fr. p-z eloszlása:

$$W_{PID}(s) = \frac{A_p}{T_I} \frac{1 + s T_I + s^2 T_I T_D}{s} \quad ; \quad z_{1,2} = \frac{-T_I \pm \sqrt{T_I^2 - 4 T_I T_D}}{2 T_I T_D}$$

$$T_I^2 - 4 T_I T_D \geq 0 \Rightarrow T_D \leq \frac{T_I}{4}$$

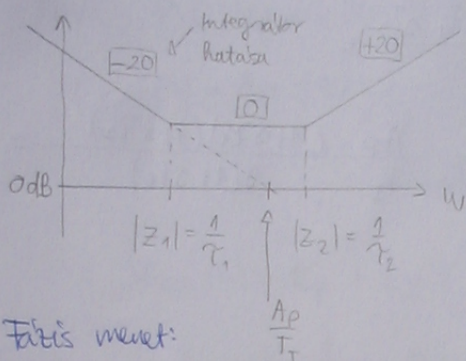
legy mindig negatívok a zelmok!



$$W_{PID}(s) = \frac{A_p}{T_I} \frac{(1 + s z_1)(1 + s z_2)}{s}$$

Bode-diagrammja:

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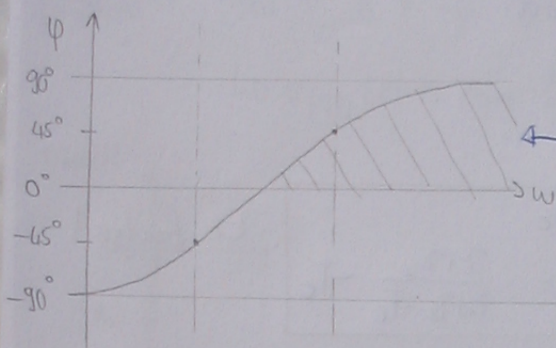


Apl. tuncó mevet

Probléma:

- klerősíti a nagy frekvenciás zajokat
- induláskor ∞ nagy jelet akar adni

Fázis mevet:



← fázis többlet jó indulni járulását okozza majd

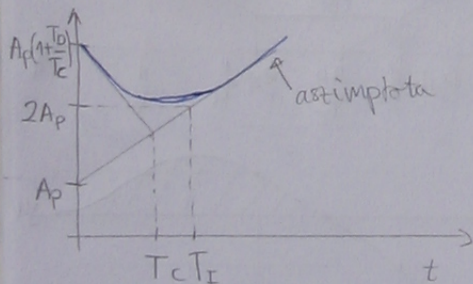
Gyakorlatban: Közelítő PID szabályzó (nem ideális)

$$W_{PID}(s) = A_p \left(1 + \frac{1}{sT_I} + \frac{sT_D}{1+sT_c} \right) = A_p + \frac{A_p}{T_I} \cdot \frac{1}{s} + \frac{A_p T_D}{T_c} \frac{s}{s + \frac{1}{T_c}}$$

$$V_{PID}(s) = W_{PID}(s) \cdot \frac{1}{s} = \frac{A_p}{s} + \frac{A_p}{T_I} \frac{1}{s^2} + \frac{A_p T_D}{T_c} \frac{1}{s + \frac{1}{T_c}}$$

$$v_{PID}(t) = A_p + \frac{A_p}{T_I} \cdot t + \frac{A_p T_D}{T_c} e^{-\frac{t}{T_c}} ; t \geq 0$$

Átmeneti fr.:



Visselkedés felvételének tartományban:

$$\begin{aligned}
 W_{PID}(s) &= \frac{A_P}{T_I} \frac{\Delta T_I (1 + sT_c) + 1 + sT_c + \Delta T_I \Delta T_D}{\Delta (1 + sT_c)} = \\
 &= \frac{A_P}{T_I} \frac{1 + \Delta(T_I + T_c) + \Delta^2 T_I (T_D + T_c)}{\Delta (1 + sT_c)} = \frac{A_P}{T_I} \frac{(1 + \Delta \tau_1)(1 + \Delta \tau_2)}{\Delta (1 + sT_c)} = \\
 &= \frac{A_P}{T_I} \frac{1 + \Delta(\tau_1 + \tau_2) + \Delta^2 \tau_1 \tau_2}{\Delta (1 + sT_c)}
 \end{aligned}$$

A két kört szimulációja egyenlő. Ezért:

$$\tau_1 + \tau_2 = T_I + T_c$$

$$\tau_1 \cdot \tau_2 = T_I (T_D + T_c)$$

$$\begin{array}{l}
 \tau_1, \tau_2, T_c \longrightarrow \\
 \boxed{
 \begin{array}{l}
 T_I = \tau_1 + \tau_2 - T_c \\
 T_D = \frac{\tau_1 \tau_2}{T_I} - T_c = \frac{\tau_1 \tau_2}{\tau_1 + \tau_2 - T_c} - T_c
 \end{array}
 }
 \end{array}$$

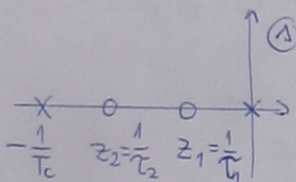
Ez fontos!

Bode:

$$z_{1,2} = \frac{-(T_I + T_c) \pm \sqrt{(T_I + T_c)^2 - 4T_I(T_D + T_c)}}{2T_I(T_D + T_c)}$$

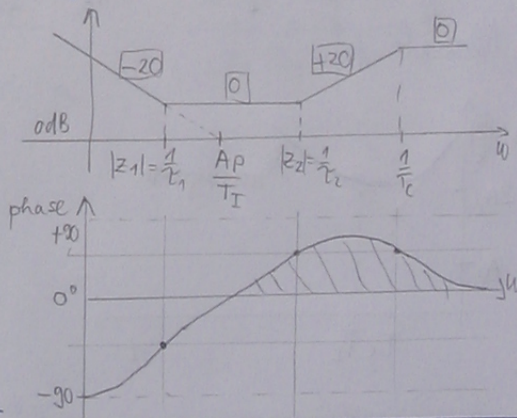
$$(T_I + T_c)^2 - 4T_I(T_D + T_c) = T_I^2 + 2T_I T_c + T_c^2 - 4T_I T_D - 4T_I T_c =$$

$$= (T_I - T_c)^2 - 4T_I T_D \Rightarrow T_D \leq \frac{(T_I - T_c)^2}{4T_I} \Rightarrow 2 \text{ valós zérushely van}$$



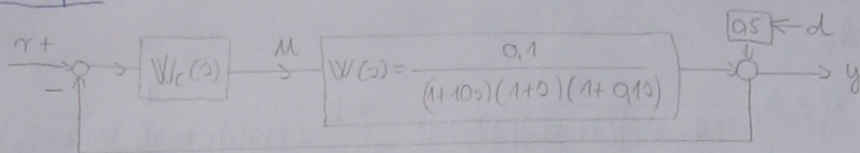
Csak ezt az elrendezést
stabilitásért használjuk!

//// - stabilizálásra
felhasználható! -18-



Minta pl:

pelda miatt a nevemben a 10-es arad! IX. sz. P s.k.



$$d(t) = d_0 1(t), \quad |d_0| \leq 10$$

$$y(\infty) = \frac{0,5 \cdot 10}{1+K} \leq 0,05 \text{ (spec.)} \Rightarrow \frac{5}{1+K} \leq 0,05$$

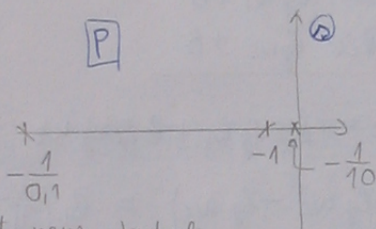
$$1+K \geq 100 \Rightarrow K \geq 99$$

Döntés: P, PD esetén: $K=100$

PI, PID -u- : K testetősleges

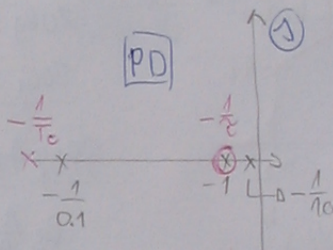
Stabilitási tartalék legyen: $\varphi_t \geq 45^\circ$

Időállandók: 0,1; 1; 10



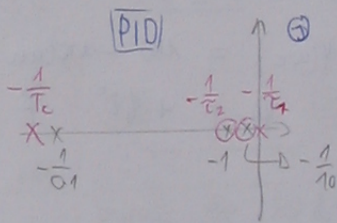
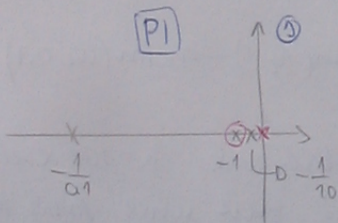
Itt nem tudunk

változtatni, csak A_p állítható!



Ezt ajánlatos kiejtetni.

2. legnagyobb időállandót abszolútú kiejtetni!



2 domináns pólust egyjűk ki!

↑ EZ a kompenzációs fő elve! ↑

P szabályozás

$$W_c(s) = A_p$$

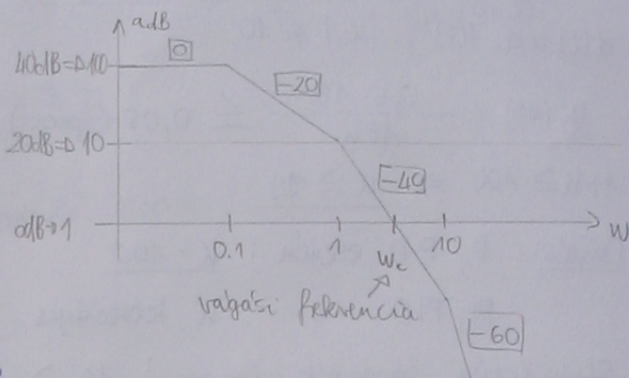
$$W_0 = W_c W = A_p \frac{A}{(1+sT_1)(1+sT_2)(sT_3+1)} = K \frac{1}{(1+sT_1)(1+sT_2)(1+sT_3)}$$

$$K = A_p A ;$$

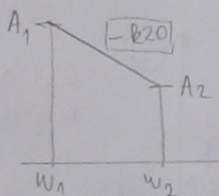
$$\Delta_1 = -\frac{1}{T_1} = -\frac{1}{10} = -0.1$$

$$\Delta_2 = -\frac{1}{T_2} = -\frac{1}{1} = -1$$

$$\Delta_3 = -\frac{1}{T_3} = -\frac{1}{0.1} = -10$$



$$20 \lg |W| = \boxed{-k20} \lg w + B$$



$$20 \lg A_1 = -k20 \lg w_1 + B$$

$$20 \lg A_2 = -k20 \lg w_2 + B$$

$$20 \lg A_1 - 20 \lg A_2 = -k20 \lg w_1 + k20 \lg w_2$$

$$\lg \frac{A_1}{A_2} = k \underbrace{(\lg w_2 - \lg w_1)}_{\lg \frac{w_2}{w_1}} = \lg \left(\frac{w_2}{w_1} \right)^k$$

$$\frac{10}{1} = \left(\frac{w_c}{1} \right)^2 \Rightarrow w_c = \sqrt{10} = 3.16$$

$$\varphi_t = 180^\circ + \varphi(w_c) = 180^\circ - \arctan(w_c \cdot 10) - \arctan(w_c \cdot 1) - \arctan(w_c \cdot 0.1)$$

$$\varphi_t = +1.8^\circ \neq 45^\circ$$

\Rightarrow P szabályozó $K=100$ esetén nem megfelelő, mert nincs fázis tart.

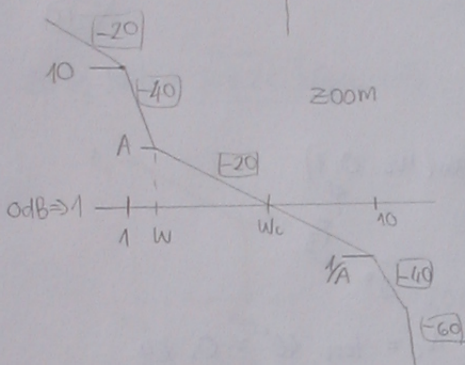
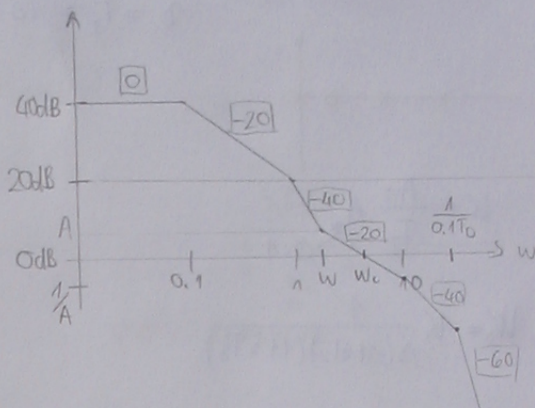
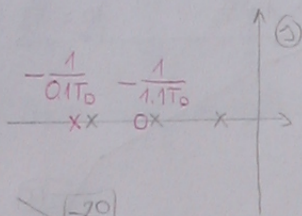
Kompensálás közelítő PD szabályzódal

IX. 28 p
3h

$$W_{PD} = A_p \left(1 + \frac{T_0}{T_c}\right) \left(1 + \frac{s T_0}{1 + s T_c}\right) = \frac{A_p (1 + s T_0 T_c)}{1 + s T_c} = A_p \frac{1 + 1.1 T_0 s}{1 + 0.1 T_0 s}$$

Ökör szabály:

Legyen: $T_c = 0.1 T_0$



$$\frac{10}{A} + \left(\frac{\omega}{1}\right)^2$$

$$\frac{A}{1/A} = \left(\frac{10}{\omega}\right)^1 \Rightarrow \omega = \frac{10}{A^2}$$

$$\frac{10}{A} = \left(\frac{10}{A^2}\right)^2 \Rightarrow A^3 = 10 \Rightarrow A = \sqrt[3]{10} = 2.15$$

$$\omega = \frac{10}{A^2} = 2.15$$

$$\frac{A}{1} = \frac{\omega_c}{\omega} \Rightarrow \omega_c = A \omega = 4.64$$

$$1.1 T_0 = \frac{1}{\omega} \Rightarrow T_0 = 0.42$$

$$T_c = 0.1 T_0 = 0.042$$

$$\varphi_t = 180^\circ - \arctan(\omega_c \cdot 10) - \arctan(\omega_c \cdot 1) + \arctan(\omega_c \cdot 1.1 T_0)$$

↑
4.64

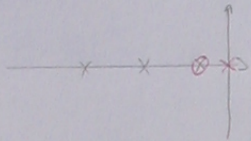
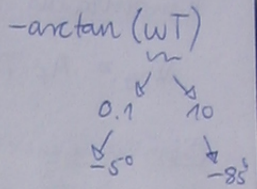
$$- \arctan(\omega_c \cdot 0.1) - \arctan(\omega_c \cdot \underbrace{0.1 T_0}_{T_c}) = 42.4^\circ = \varphi_t \approx 45^\circ$$

Ez már jó!

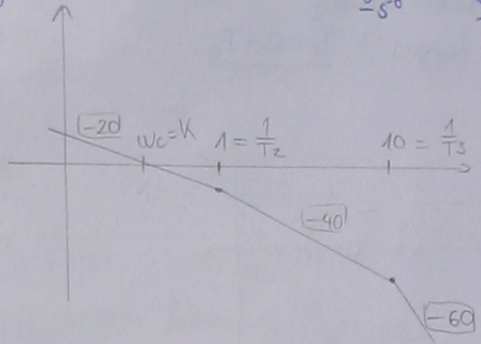
P1 kompenzálás

$$W_{PI} = \frac{A_P}{T_I} \frac{1+sT_I}{s} = \frac{A_P}{T_I} \frac{1+s\tau}{s}$$

$$\tau = T_I = 10$$



$$K = \frac{A_P}{T_I} A \leftarrow 0.1$$



$$W_0 = K \frac{1}{s(1+sT_2)(1+sT_3)}$$

$$\varphi_t = 180^\circ + \varphi(\omega_c) =$$

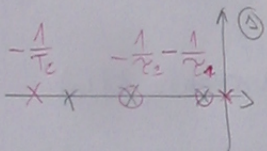
$$45^\circ = 180^\circ - 90^\circ - \underbrace{\arctan(\omega_c \cdot \frac{1}{T_2}) - \arctan(\omega_c \cdot 0.1)}_{-5^\circ}$$

$$\frac{90^\circ - 45^\circ - 5^\circ}{40^\circ} = \arctan(\omega_c \cdot 1) \Rightarrow \omega_c = \tan 40^\circ = 0.84$$

$$\omega_c = K = \frac{A_P}{T_I} \cdot A = \frac{A_P}{10} \cdot 0.1 \Rightarrow \boxed{A_P = 84}$$

$$\boxed{T_I = 10 \text{ sec}}$$

PID stabilizálás



$$T_c = 0.1 T_D$$

$$W_{PID} = \frac{A_P}{T_I} \frac{1+s(T_I + 0.1T_D) + s^2 T_I T_D}{s(1+s \cdot 0.1 T_D)}$$

$$W_0 = \frac{A_P}{T_I} \frac{(1+s\tau_1)(1+s\tau_2)}{s(1+s \cdot 0.1 T_D)} \cdot \frac{A}{(1+sT_1)(1+sT_2)(1+sT_3)}$$

$$\tau_1 = T_1, \tau_2 = T_2 \Rightarrow (1+s\tau_1)(1+sT_2) = 1+s(\tau_1+T_2) + s^2\tau_1 T_2 =$$

$$= 1 + 11s + 10s^2$$

$$T_I + 0,1 T_D = 11$$

$$T_I \cdot 1,1 T_D = 10 \rightarrow T_I = \frac{10}{1,1 T_D}$$

$$\frac{10}{1,1 T_D} + 0,1 T_D = 11 \Rightarrow 10 + \underbrace{0,1 \cdot 1,1 T_D^2}_{0,11} = \underbrace{11 \cdot 1,1 T_D}_{12,1}$$

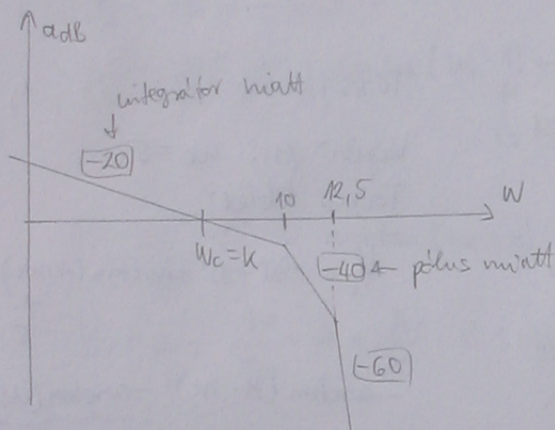
$$T_D^2 - 110 T_D + 90,9 = 0$$

$$T_D \rightarrow 109,16$$

$$T_D \rightarrow 0,83 \rightarrow T_C = 0,1 T_D = 0,083 \Rightarrow T_I = \frac{10}{1,1 \cdot 0,83} = 10,92$$

$$W_0 = \frac{K}{\Delta (1 + 0,1) (1 + 0,083)}$$

\uparrow \uparrow
 T_D T_C



$$45^\circ = 180^\circ - 90^\circ - \arctan(W_c \cdot 0,1) - \arctan(W_c \cdot 0,083)$$

$$45^\circ = \arctan(W_c \cdot 0,1) + \arctan(W_c \cdot 0,083)$$

$$\tan(\alpha + \beta) = \frac{\tan(\alpha) + \tan(\beta)}{1 - \tan(\alpha)\tan(\beta)} \quad ; \quad \tan(\arctan(x)) = x$$

$$1 = \frac{W_c \cdot 0,1 + W_c \cdot 0,083}{1 - W_c \cdot 0,1 \cdot W_c \cdot 0,083} \Rightarrow W_c^2 + 22 W_c - 120,5 = 0 \Rightarrow W_c = 4,55$$

$$K = \frac{A_P}{T_I} \quad A = \frac{A_P}{10,92} \quad 0,1 = W_c = 4,55 \Rightarrow$$

$A_P = 4,97$
$T_I = 10,92 \text{ sec}$
$T_D = 0,83 \text{ sec}$
$T_C = 0,083 \text{ sec}$

Kiegészítés (Extended version)

Szükség a 2. gyakorlat ellenőrző leckeire!

②

$$W_0 = \frac{25(\Delta + 0.1)}{\Delta(\Delta + 1)(\Delta + 5)}$$

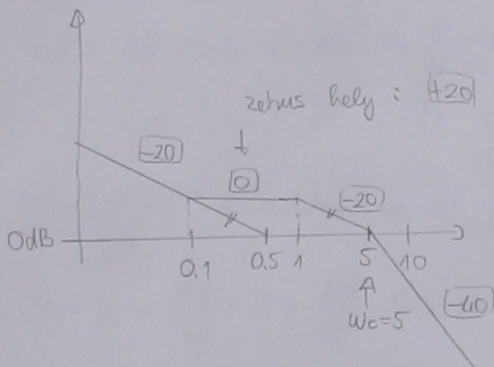
Éz gyökfőnyelés, ezert alt fell vmi.

$$W_0 = \frac{25 \cdot 0.1}{1 \cdot 5} \cdot \frac{1 + 10\Delta}{\Delta(\Delta + 1)(0.2\Delta + 1)} = 0.5 \frac{1 + 10\Delta}{\Delta(1 + \Delta)(1 + 0.2\Delta)}$$

$$= \frac{K}{s} W_{01}(s) \quad , \quad W_{01}(1)$$

$K=0.5$; típus szám = 1 (integráló elem)

↳ körszámok



Törés frekvenciák: 0,1; 1; 5

Válasz fri: $w_c = 5$

Fázis többlet!

$$\varphi_t = 180^\circ - 90^\circ + \arctan(10w_c)$$

$$- \arctan(1 \cdot w_c) - \underbrace{\arctan(0.2w_c)}_{-45^\circ}$$

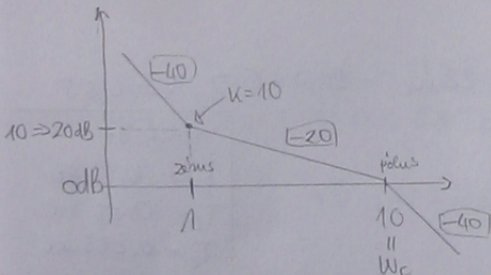
Ht a törtet: $(1 + Ts)(\dots)$ alakra kellett hozni!

④

$$W_0 = \frac{10(1 + s)}{s^2(1 + 0.1s)} = \frac{K}{s^2} W_{01}(s) \quad , \quad W_{01}(0) = 1$$

$K=10$, típus szám = 2 (1/n hatványa)

$$|W_0| \sim \frac{K}{w^2} \Rightarrow w=1 \text{ -nél: } |W_0| = K$$



$w_c = 10$

$$\varphi_t = 180^\circ - 90^\circ - 90^\circ + \arctan(w_c \cdot 1)$$

$$\arctan(w_c \cdot 0.1) \cong$$

$$\cong 85^\circ - 45^\circ - 40^\circ$$